



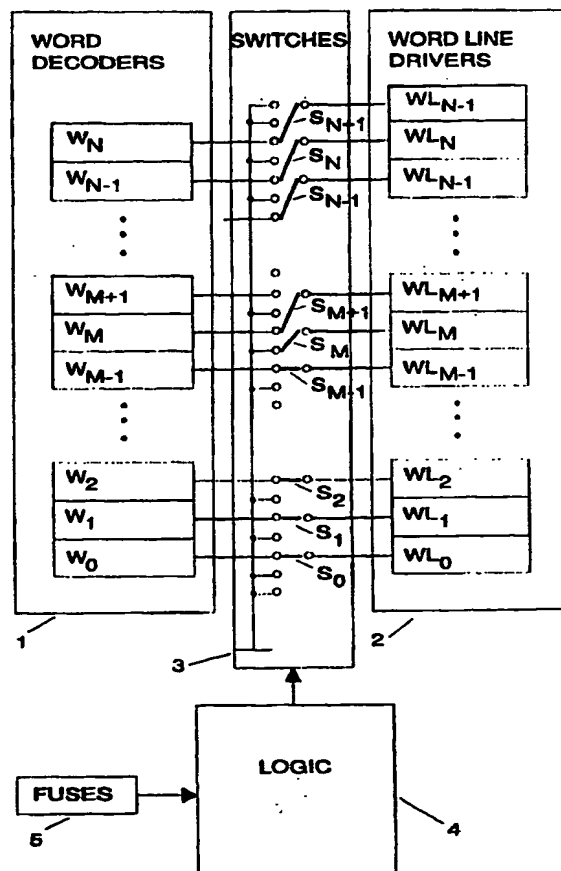
## INTERNATIONAL APPLICATION PUBLISHED UNDER THE PATENT COOPERATION TREATY (PCT)

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(54) Title: STATIC WORDLINE REDUNDANCY MEMORY DEVICE

## (57) Abstract

The invention relates to a memory device comprising a set of word decoders (W), a set of wordline drivers (WL), a plurality of switches (S) to connect a subset of the wordline drivers to the set of word decoders and storage means (5) for storage of information indicative of defective wordline, the wordline drivers comprising a predefined first subset of wordline drivers which are to be used when no wordline is defective, the set of wordline drivers further comprising a plurality of second subsets of wordline drivers which are to be used when one of the wordlines is defective, the memory device further comprising logic means (4) for logically and permanently assigning one of the subsets to the set of word decoders in response to the information stored in the storage means, the logic means (4) controlling the switches (S) to connect the second subset of wordline drivers to the set of word decoders.



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## D E S C R I P T I O N

## STATIC WORDLINE REDUNDANCY MEMORY DEVICE

The present invention relates to a memory device and method implementing wordline redundancy without an access time penalty.

The application of wordline redundancy to enhance the yield for memory arrays is an accepted fact throughout the semiconductor industry. To be attractive, wordline redundancy should occur without major impact to chip performance (e.g. access time), power requirements or size. Numerous approaches have been proposed with varying degrees of success; for example:

U.S. Patent No. 4,365,319, issued to Takemae on December 21, 1982, implements redundancy by utilizing two kinds of decoders and drivers, i.e., a PROM decoder for determining whether an incoming address is a defective address, a redundancy driver for driving a redundancy array, and row address decoders and drivers for driving a main memory cell matrix. A first embodiment of the Takemae teachings (Fig. 1) is disadvantageous in that the switch 7 results in an access time penalty and results in a semiconductor space penalty because the switch must be large to handle high currents. In a second embodiment (Figs. 2-4), multiple AND gates  $D_0$ - $D_{63}$  replace the large switch 7 (Fig. 1); however, this is not much of an improvement because the memory device still suffers from both an access time (e.e., an AND-gate) penalty, and also a semiconductor space penalty as the collective area of the AND gates  $D_0$ - $D_{63}$  is still large. A third embodiment (Figs. 5-10) suffers an access time penalty due to AND-gate delays introduced by the incorporation of AND gates  $D_{91}$ - $D_{94}$  and (Fig. 6) and AND gates  $D_0$ - $D_3$  (Fig. 8A) to control the activation of the decoders and drivers 9 and

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10, irrespectively.

U.S. Patent No. 3,753,244, issued to Sumilas et al on August 14, 1973, implements redundancy by placing an extra line of memory cells on a memory chip together with a defective address store and a comparator circuit for disabling a defective line of cells and replacing it with the extra line of cells.

The Intel 2164A 64K DRAM represents a memory device where access time is the same whether it is the normal wordlines or the redundancy wordlines which are being used; however, this product is always affected by an access time penalty, whether repaired with wordline redundancy or not, because chip timing is set up to allow for redundancy repairs. More specifically, chip performance is slowed due to the need to deselect a faulty wordline's word decoder after the redundant word decoders sense a match with an incoming address. Once the match is sensed, a deselect generator is fired to deselect the entire row of normal word decoders. After the faulty wordline word decoder is deselected, then the wordline drive is enabled. Further discussions concerning the 2164A can be seen in the Intel Application Description AP-131, pp. 14-16, and "An Analysis of the i2164A", Mosaid Incorporated, p. 5, 41-52, April 1982. In addition, it should be further noted that IBM has a 72k DRAM which utilizes a similar approach.

The Bell Lab 64K DRAM (described by R.T. Smith, J.D. Chilipala, J.F.M. Bindels, R.G. Nelson, F.H. Fischer And T.F. Mantz, in "Laser Programmable Redundancy and Yield Improvement in a 64K DRAM", IEEE Journal of Solid-State Circuits, Vol. SC-16, No. 5, pp. 506-514, October 1981), and the 256K DRAM (described by C.A. Benevit, J.M. Cassard, K.J. Dimmler, A.C. Dumbri, M.G. Mound, F.J. Procyk, W.R. Rosenzweig and A.W. Yanof, in "A 256k

Dynamic Random Access Memory", IEEE Journal of Solid-State Circuits, Vol. SC-17, No. 5, pp. 857-861, October 1982), implement wordline redundancy without an access time impact by using laser-fused redundancy on the wordline pitch. No access time penalty is incurred because the defective wordline is permanently disconnected by exploding a programmable link in the wordline. This method of redundancy is disadvantageous because the tighter design rules of present and future high density memory products are causing a shrinkage in the wordline pitch. the result is a requirement for a laser spot size and laser beam position accuracy beyond what is available from laser programming systems today. Thus, laser-fused redundancy is disadvantageous in that the current level of laser technology requires an off wordline pitch method or an increase in memory chip size due to the need for an increased wordline pitch.

The IBM 32K DRAM (described by B.F. Fitzgerald and E.P. Thoma, in "Circuit Implementation of Fusible Redundant Addresses on RAMs for Productivity Enhancement", IBM Journal of Research and Development, Vol. 24, No. 3, pp. 291-295, May 1980) implements wordline redundancy without an access time penalty by adding separate sense amplifier columns for the redundant wordlines. No access penalty is incurred because the redundant wordline and the defective wordline operate in parallel, and the selection of the redundant, versus the normal sense amplifiers, occurs during the sensing operation. This approach is disadvantageous in that chip size is significantly increased due to the need for additional latches for each bitline along the redundant wordline.

Similarly, R.P. Cenker, D.G. Clemons, W.R. Huber, J.B. Petrizzi, F.J. Procyk and G.M. Trout, in "A Fault-Tolerant 64K Dynamic Random Access Memory", IEEE Transactions on Electron Devices, Vol. ED-26, No. 6, June

1979, teach a word redundancy technique having no access time penalty, but requiring that disabling fuses be placed within each redundant and non-redundant decoder, thus significantly increasing the required chip area.

B.F. Fitzgerald and D.W. Kemerer, in "Memory System With High-Performance Word Redundancy", IBM Technical Disclosure Bulletin, Vol. 19, No. 5, October 1976, describe an implementation of word redundancy with no access penalty by accessing both a normal and redundant row in independent arrays. Selection of good data was made at the data out buffers.

From EP-A-0 336 101 a semiconductor memory device and method for implementing wordline redundancy is known. A redundant word decoder compares an incoming address signal with a list of defective addresses and, in response to the comparison, produces at least one comparison signal to control the propagation of a redundant driver signal along at least one redundant wordline. A main trigger receives the comparison signal and, in response thereto, triggers the firing of a main wordline driver to produce a main driver signal. The main wordline driver and the redundant word decoder are responsive to opposite states of the comparison signal, such that for a given comparison signal, only one of the main driver signal and redundant driver signal is applied to a memory array.

From EP-A-0 029 322 a semiconductor device, in which a redundancy memory cell array is incorporated with a main memory cell matrix, is known. One memory cell array is selected by two kinds of decoders and drivers. When the redundancy memory cell array is selected by a decoder, the decoder disables one kind of the decoders and drivers directly and, as a result, the other kind of the decoders and drivers are also disabled.

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A semiconductor memory device wherein a redundancy memory cell array incorporated with main memory cell matrixes is disclosed in US-A-4,392,211. Memory cells of the main memory cell matrixes are selected by first and third decoders while memory cells of the redundancy memory cell array are selected by second and third decoders. When the redundancy memory cell array is selected by the second decoder, the transmission of a clock signal to the first decoders is stopped by a switching circuit.

While the above approaches represent important advances in semiconductor manufacturing technology, there still exists a need for an improved memory device and approach which are able to provide wordline redundancy. It is therefore an object of the invention to provide an improved memory device and method to implement wordline redundancy.

The object of the invention is solved by the features laid down in the independent claims.

The inventive memory device comprises a set of word decoders and a set of wordline drivers. The number of the wordline drivers is greater than the number of the word decoders. This implies that the physical real address space is larger than the addressable address space because each wordline driver is connected to a different wordline. If one or more of the wordlines are defective a subset of the wordline drivers is selected which does not comprise the wordline drivers belonging to the defective wordlines. This set of wordline drivers is different from the normal set of wordline drivers which is used when no wordline is defective.

The memory device further comprises a storage means for storage of information indicative of a defective wordline. This can be realized by a "Fuse address". Once

the memory device has a power supply voltage applied thereto such a subset of wordline drivers is selected by a logic means according to the information indicative of a defective wordline. The subset which is selected by the logic means is permanently assigned to the set of word decoders. Furthermore the logic means control the switches between the word decoders and wordline drivers to connect the subset of wordline drivers which is selected by the logic means to the set of word decoders. Thereby each wordline driver of the selected subset of wordline drivers is permanently connected to a specific one of the word decoders.

The selection and connection of the wordline drivers is already accomplished before the memory device is operated, e.g. to read and write data. Once the permanent connection of the wordline drivers is established no further steps are necessary in order to implement the wordline redundancy because the connections of the wordline drivers to the word decoders are static. Hence, no further decoding or switching operations have to take place when the memory device is actually operated to read or write data.

In principle, the number of redundant wordlines is not restricted by the invention. If, for example, there is only one redundant wordline, this requires also one additional switch. Therefore a number  $n$  of word decoders requires a number  $n+1$  of wordline drivers of the  $n+1$  wordlines and a number  $n+1$  of switches because each wordline driver requires one switch.

In the example considered here the logic means has to generate 3 possible control states for each of the switches: a first control state indicates that the corresponding switch has to be connected to its "normal" word decoder to which the switch is also connected when



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there is no defective wordline.

A second control state indicates that the corresponding switch should disconnect its associated wordline driver from the word decoders, because the wordline driver belongs to a defective wordline and is to be replaced by another wordline driver. Thereby the wordline driver belonging to a defective wordline is disabled. This can be accomplished by grounding the wordline driver.

The 3rd state of the logic means indicates that the corresponding switch does not have to connect its wordline driver to the "normal" word decoder to which the wordline driver is connected when there is no defective wordline. In this case the wordline driver is connected to another word decoder which is not already connected to a wordline driver via a switch being in the first control state. This can be for example the word decoder which precedes the "normal" word decoder to which the wordline driver is connected when there is no defective wordline. This principle of operation can analogously be realized for two or more redundant wordlines.

A computer system incorporating a memory device according to the invention features improved speed of operation as compared to the prior art because the implementation of wordline redundancy results in no performance penalty. Furthermore, the invention is advantageous in that the realization of the inventive principle does only require relatively few electrical components and - as a consequence - requires only relatively little space on the chip.

In the following one way of carrying out the invention is shown in more detail with reference to the drawings in which:

- Fig. 1      schematically shows the connection of the word decoders and the wordline drivers via the switches;
- Fig. 2      schematically shows the realization of the logic means comprising a plurality of logic blocks;
- Fig. 3      is a circuit diagram showing in more detail the realization of one of the logic blocks;
- Fig. 4      shows in more detail the realization of the decoder which is incorporated in the logic block;
- Fig. 5      is a circuit diagram showing the realization of one of the switches.

As it is shown in Fig. 1 a set of word decoders 1 is connected to a subset of the set of wordline drivers 2 via a plurality of switches 3. In the example considered here the set of word decoders comprises the word decoders  $W_0, W_1, W_2, \dots, W_{m-1}, W_m, W_{m+1}, \dots, W_{n-1}, W_n$ . The set of wordline drivers 2 comprises the wordline drivers  $WL_0, WL_1, WL_2, \dots, WL_{m-1}, WL_m, WL_{m+1}, \dots, WL_{n-1}, WL_n, WL_{n+1}$ . Each of the wordline drivers  $WL$  of the set of wordline drivers 2 is connected to one wordline. The wordlines are not shown in the drawing. Since the number of wordline drivers  $WL$  is greater than the number of word decoders the physical addressed space is larger than the addressable address space. In the case considered here there is one more wordline driver than there are word decoders.

Each wordline driver  $WL$  has one of the switches of the plurality of switches 3 associated thereto. The switch  $S_0$  of the plurality of switches 3 is connected to  $WL_0$ ,  $S_1$  to

WL1, S2 to WL2, ..., Sm-1 to WLn-1, Sm to WLn, Sm+1 to WLn+1, ..., Sn-1 to WLn-1, Sn to WLn and Sn+1 to WLn+1. The number of switches S is equal to the number of wordline drivers WL.

In the example considered here the wordline WLn is assumed to be defective. As a consequence the switch Sm of the wordline driver WLn connects the wordline WLn to ground - or with other words - the switch Sm disconnects the defective wordline driver WLn from the set of word decoders 1 and thus disables the wordline driver WLn.

This situation is different from the normal situation when there is no defective wordline driver. In the normal case each word decoder of the set of word decoders 1 is connected to a wordline driver of a predefined first subset of the set of wordline drivers 2. In this example the predefined first subset of wordline drivers for the normal case is the set of wordline drivers WL0, WL1, WL2, ..., WLn-1, WLn. Hence, in the normal operation mode the word decoder W0 is connected to the wordline driver WL0, W1 to WL1, W2 to WL2, ... Wm-1 to WLn-1, Wm to WLn, Wm+1 to WLn+1, ..., Wn-1 to WLn-1 and Wn to WLn. The wordline driver WLn+1 is connected to ground by its switch Sn+1 and is thus disabled.

The situation shown in Fig. 1 is not the normal situation when there is no defective wordline driver. Since in the case shown in Fig. 1 one of the wordline drivers - in this example WLn - is defective the addressable addressed space has to be distributed differently in the physical address space as compared to the normal situation. This is accomplished by connecting the word decoders W of the set of word decoders 1 to a second subset of wordline drivers of the set of wordline drivers 2. The second subset consists of the entire set of wordline drivers 2 except the defective wordline driver WLn.

The word decoders  $W_0$  to  $W_{m-1}$  are connected to the respective wordline drivers  $WL_0$  to  $WL_{m-1}$  like in the normal situation when there is no defective wordline driver. As opposed to this the word decoders  $W_m$  to  $W_n$  are connected to the wordline drivers  $WL_{m+1}$  to  $WL_{n+1}$ . This is because the wordline  $WL_m$  is defective and disabled by the switch  $S_m$ . The wordline driver  $WL_{n+1}$  is no longer disabled but connected to the word decoder  $W_n$  via the switch  $S_{n+1}$ . Thereby the functionality of the defective wordline driver  $WL_m$  is replaced.

The memory device shown in Fig. 1 further comprises logic means 4 which is connected to the plurality of switches 3 to control each of the switches  $S_0$  to  $S_{n+1}$ . The control logic 4 is connected to a storage device 5. If there is a defective wordline the storage device 5 has the address of the defective wordline and of the corresponding wordline driver stored therein. In the example considered here the address  $A_m$  of the wordline  $m$  and thus the address of the wordline driver  $WL_m$  is stored in the storage device 5. The storage device 5 can be realized by a number of fuses which are programmed after the testing of the memory device.

Fig. 2 shows an overview of one realization of the control logic 4. The control logic 4 comprises a plurality of address space distribution logic blocks 5, 6, 7, ... For each switch  $S$  of the plurality of switches 3 there is one such address space distribution logic block (ASDL). The logic block 5 (ASDL0) belongs to the switch  $S_0$ , the logic block 6 (ASDL1) to  $S_1$  and the logic block 7 (ASDL2) to  $S_2$ . The further logic blocks ASDL3 to ASDL $_{n+1}$  which belong to the switches  $S_3$  to  $S_{n+1}$ , respectively, are not shown in Fig. 2. Each of the logic blocks has an input FUSADR which is connected to the storage device 5 for inputting of the address  $A_m$ . Furthermore each of the logic blocks has a decoder 8. The

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decoder 8 issues a signal if the address Am corresponds to the address of the wordline to which the wordline driver of the switch to which the logic block belongs matches. This results in two output signals S0 and S1 per logic block. The switch S0 is controlled by the output signals of its logic block 5 (ASDL0) S0\_0 and S1\_0. Analogously the switches S1 and S2 are controlled by the output signals S0\_1, S1\_1 and S0\_2, S1\_2, respectively. The further output signals S0\_3, S1\_3 to S0\_n+1 to S1\_n+1 are not shown in Fig. 2.

If the signal S0\_x equals logically 1 and the signal S1\_x equals logically 0 this means that the corresponding wordline driver WLx has to be connected by the switch Sx to the normal word decoder Wx. If both of the signals S0\_x and S1\_x equal logically 0 the switch Sx is controlled to disable the wordline driver WLx. If the signal S0\_x equals logically 0 and the signal S1\_x equals logically 1 the switch Sx is controlled to connect the wordline driver WLx to the word decoder Wx-1.

The logic block 5 has a further input signal FUSE\_ENB applied thereto. The input signal FUSE\_ENB is logically 1 when there is a defective wordline. In the opposite case FUSE\_ENB is logically 0. If FUSE\_ENB is logically 0 this signal passes through the AND gate 9 to the corresponding input SOEN of the next logic block 6. As a consequence the input signal FUSE\_ENB propagates through all of the logic blocks.

By way of example Fig. 3 shows one of the logic blocks, logic block 5, in more detail. It is to be understood that the circuit diagram for all of the logic blocks is identical.

The logic block 5 comprises the decoder 8, the AND gate 9 and an inverter 10. The input signal FUSADR is inputted

to the decoder 8. If the address stored in storage device 5 and thus the signal FUSADR matches the address  $A_m$  of the wordline  $m$  to which a logic block ASDL $m$  belongs the decoder 8 will issue a signal HIT\_MISS. In the case of a match of the addresses the signal HIT\_MISS is logically 0.

For the logic block 5 the signal HIT\_MISS is logically 0 when the address stored in the storage device 5 is the address A0 of the wordline 0. The signal HIT\_MISS is inputted to the AND gate 9 as well as the further input signal S0IN. In the case of the logic block 5 the input signal S0IN is the signal FUSE\_ENB. The fuse enable signal FUSE\_ENB is logically 0 if there is no defective wordline. In this case the output of AND gate 9 and thus S0\_0 is always logically 0 irrespective of the condition of the signal HIT\_MISS.

If the signal FUS\_ENB is logically 1 this indicates that there is a defective wordline. In this case the output of the AND gate 9 depends on the signal HIT\_MISS. An inverter 10 is connected to the input S0IN to produce the output S1\_0.

In the example considered here there is an address space of 5 bits. Correspondingly the decoder 8 has a NAND gate having 5 inputs A0, A1, A2, A3 and A4. The signal FUSADR comprises the address bits B0 to B4 and the complements of the address bits  $\overline{B0}$  to  $\overline{B4}$ . Whether the true or complement bits of FUSADR are connected to one of the inputs of the NAND gate of the decoder 8 depends on the address  $A_m$  to which the logic block to which the decoder 8 belongs is assigned.

This is explained in more detail with reference to Fig. 4. The first row in Fig. 4 indicates the bit positions of the signal FUSADR, i.e. B0 to B4 and  $\overline{B0}$  to  $\overline{B4}$ . The second

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row in Fig. 4 indicates which of the inputs A0 to A4 of the NAND-gate of the logic block 5 (ASDL0) are connected to which bits of the input signal FUSADR. In the ASDL0 only the complement bits  $\overline{B0}$  to  $\overline{B4}$  are used.  $\overline{B0}$  is connected to A0,  $\overline{B1}$  to A1,  $\overline{B2}$  to A2,  $\overline{B3}$  to A3 and  $\overline{B4}$  to A4.

If one assumes that the wordline 0 having the address 00000 is defective this results in an input 11111 to the NAND gate of the decoder 8 of the ASDL0. Hence the signal HIT\_MISS of the ASDL0 is logically 0 which indicates that an address match occurs. Analogously the bit B0 of FUSADR is connected to the input A0 of the NAND-gate of the decoder 8 of the ASDL1 whereas the inputs of A1 to A4 remain unchanged. The same principle applies to the connection of the bit positions of the signal FUSADR to the further logic blocks ASDL2, ASDL3, ..., ASDLn+1.

Fig. 5 shows one implementation of a switch S. By way of example the switch 11 shown in Fig. 5 is considered to be the switch Sm+1. The switch Sm+1 has inputs 12 and 13 which are connected to the word decoders Wm+1 and Wm. Furthermore the switch Sm+1 is connected to the signals S0\_m+1 and S1\_m+1 of its ASDLm+1 at inputs 14 and 15. The output 16 of the switch Sm+1 is connected to the wordline driver WLm+1 of this switch. The switch Sm+1 serves to selectively establish a connection between the word decoders Wm+1 or Wm and the wordline driver WLm+1 depending on the conditions of the control signals S0\_m+1 and S1\_m+1. If the wordline driver WLm+1 belongs to a defective wordline m+1 the switch Sm+1 is to disable the wordline driver WLm+1. This is accomplished by means of the internal circuit of the switch Sm+1 as shown in Fig. 5.

The control signals S0\_m+1 and S1\_m+1 are connected to the NOR-gate 12. The output of the NOR-gate 12 is

connected to the base of a transistor 13. One terminal of the transistor 13 is connected to the output 16 whereas the other terminal of the transistor 13 is connected to ground. If an address match of the address stored in the storage device 5 and the wordline  $m+1$  occurs this results in the control signals  $S0_{m+1}$  and  $S1_{m+1}$  to both equal logically 0 (cf. Fig.3 and Fig.4). As a consequence the output of the NOR-gate 12 is logically 1 so that the transistor 13 connects the output 16 of the switch  $Sm+1$  to ground. As a consequence the wordline driver  $Wm+1$  is disconnected from the word decoders and is disabled.

Furthermore the switch  $Sm+1$  has pass gates 17 and 18. The pass gate 17 has one terminal connected to the input 12 and thus to the word decoder  $Wm+1$ . The other terminal of the pass gate 17 is connected to the output 16 and thus to the wordline driver  $Wm+1$ . The pass gates 17 and 18 consist of two complementary transistors according to the CMOS technology employed for the realization of this preferred embodiment. The gates of the transistors of the pass gate 17 are connected to the input 14 whereby the P-type transistor of the pass gate 17 has an inverter interconnected in the signal path. The same applies analogously to the pass gate 18. The gates of the pass gate 18 are connected to the input 15 and thus to the control signal  $S1_{m+1}$ . If both of the control signals equal logically 0 both of the pass gates 17 and 18 are not conductive so that no connection is established between the word decoders and the wordline driver  $Wm+1$ . If the control signal  $S0_{m+1}$  is logically 1 the word decoder  $Wm+1$  is connected to the wordline driver  $Wm+1$ . In this case the control signal  $S1_{m+1}$  is logically 0 since only one of the word decoders can be connected to the wordline driver  $Wm+1$  at a time.

If the control signal  $S0_{m+1}$  is logically 0 and the control signal  $S1_{m+1}$  is logically 1 this results in the



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word decoder  $W_m$  to be connected to the wordline driver  $W_{Lm+1}$ . This situation corresponds to the case shown in Fig. 1.

The switching operations of the switches take place already when a power supply voltage is applied to the memory device. Once the connections between word decoders and wordline drivers have been established via the switches under the control of the control logic 4 these connections remain unchanged - at least as long as a power supply voltage is applied to the memory device. As a consequence there is no access time penalty since no switching or decoding operation has to be carried out "On the fly" when the memory device is used to carry out read/write operations. The information that a wordline is defective is stored on the memory device after the testing of the device in order to program the signal `FUSE_ENB` as well as the address of the defective wordline in order to program the signal `FUSADR`.

## C L A I M S

1. Memory device comprising a set of word decoders (1), a set of wordline drivers (2), a plurality of switches (3) to connect a subset of said wordline drivers to said set of word decoders and storage means (5) for storage of information indicative of a defective word line,

said set of wordline drivers comprising a predefined first subset of wordline drivers which are to be used when no wordline is defective,

said set of wordline drivers comprising a plurality of second subsets of wordline drivers which are to be used when one of said wordlines is defective,

said memory device further comprising logic means (4) for assigning one of said subsets to said set of word decoders in response to the information stored in said storage means, said logic means controlling said switches correspondingly.

2. Memory device according to claim 1 comprising a number  $n$  of word decoders, a number  $n+1$  of wordline drivers and a number  $n+1$  of switches  $S$ , wherein said switches  $S$  take the following states in response to the control of said logic means:

a) if none of said wordlines is defective:

each switch  $S_i$  connecting a wordline driver  $i$  to a corresponding word decoder  $i$ , where  $0 < i < n$ , the switch  $S_{n+1}$  disconnecting the wordline driver  $n+1$ ;

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b) if a wordline  $m$  of said wordlines is defective:

each switch  $S_i$  connecting a wordline driver  $i$  to a corresponding word decoder  $i$ , where  $0 \leq i < m$ ;

a switch  $S_m$  disconnecting said wordline driver  $m$ ;

each switch  $S_i$  connecting a wordline driver  $i+1$  to a corresponding word decoder  $i$ , where  $m < i \leq n+1$ .

3. Memory device according to claim 1 or 2 said logic means comprising a plurality of  $n+1$  logic blocks (ASDL), wherein each of said logic blocks is assigned to one of said switches  $S$  and wherein each of said logic blocks is adapted to generate a control signal for the switch  $S_i$  to which said logic block is assigned.
4. Memory device according to claim 3 each of said logic blocks having an input for inputting the address  $A_m$  of a defective wordline  $m$  and having a decoding means (8) for issuing a signal (HIT\_MISS) if said address  $A_m$  corresponds to the wordline driver  $m$  to which said logic block is assigned.
5. Memory device according to claim 3 or 4 each of said logic blocks being adapted to generate two control signals  $S_0$  and  $S_1$  for the switch  $S$  to which said logic block is assigned, each of said logic blocks having an enable input (SOIN), only one of said blocks being adapted to be connected to an enable signal (FUSE\_ENB), whereas the remaining blocks are adapted to receive one of said control signals of another one of said logic blocks at their respective

enable inputs.

6. Memory device according to any one of the preceding claims said storage means being a ROM, such as fuses.
7. Integrated circuit chip incorporating a memory device according to any one of the preceding claims.
8. Computersystem comprising a memory device according to any one of the claims 1 to 6.
9. A method to realize wordline redundancy in a memory device according to any one of the preceding claims comprising the steps of
  - a) storing information indicative of a defective wordline if at least one of said wordlines is defective;
  - b) permanently connecting a subset of said wordline drivers to said set of word decoders.

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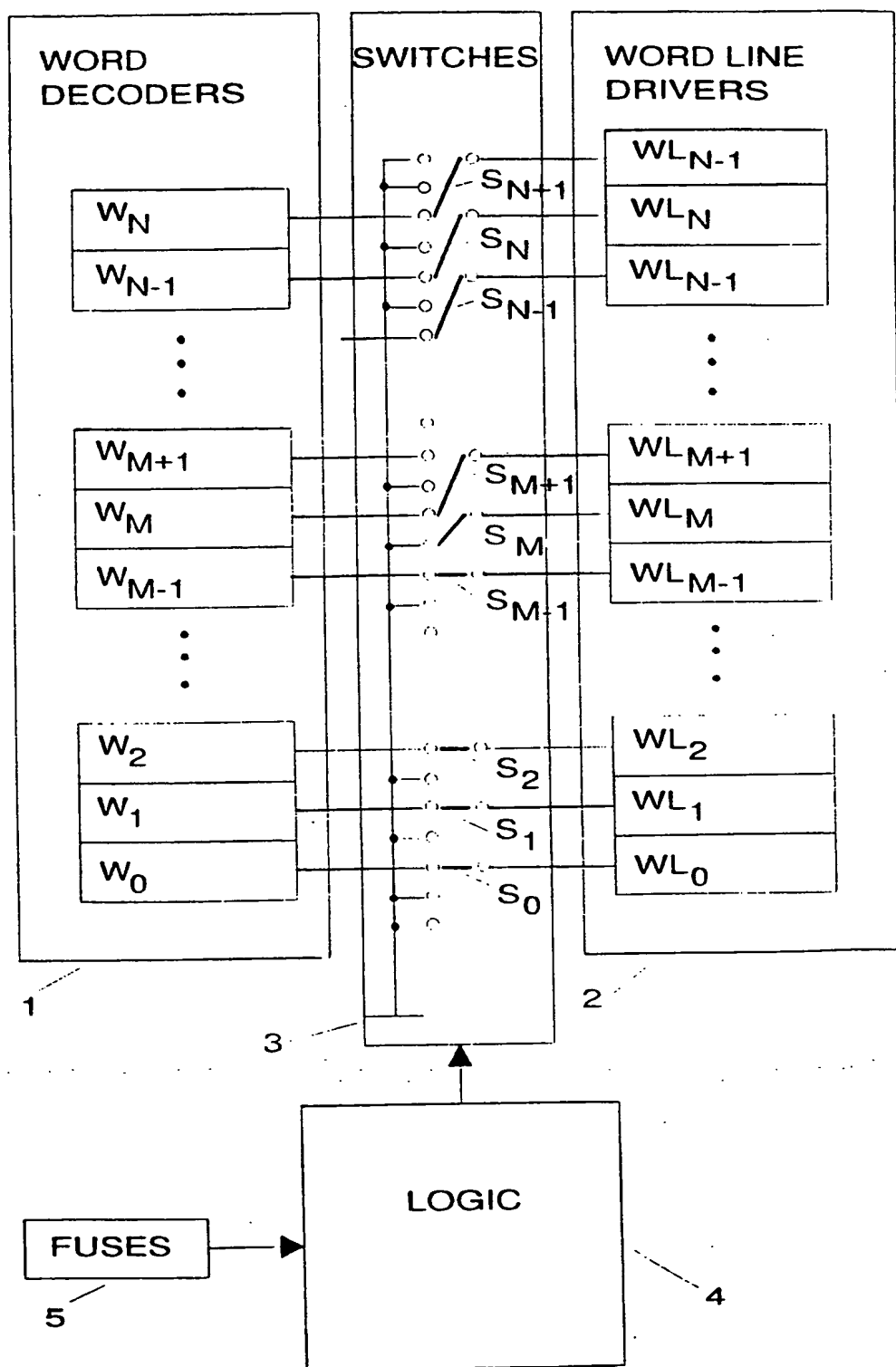


FIG. 1

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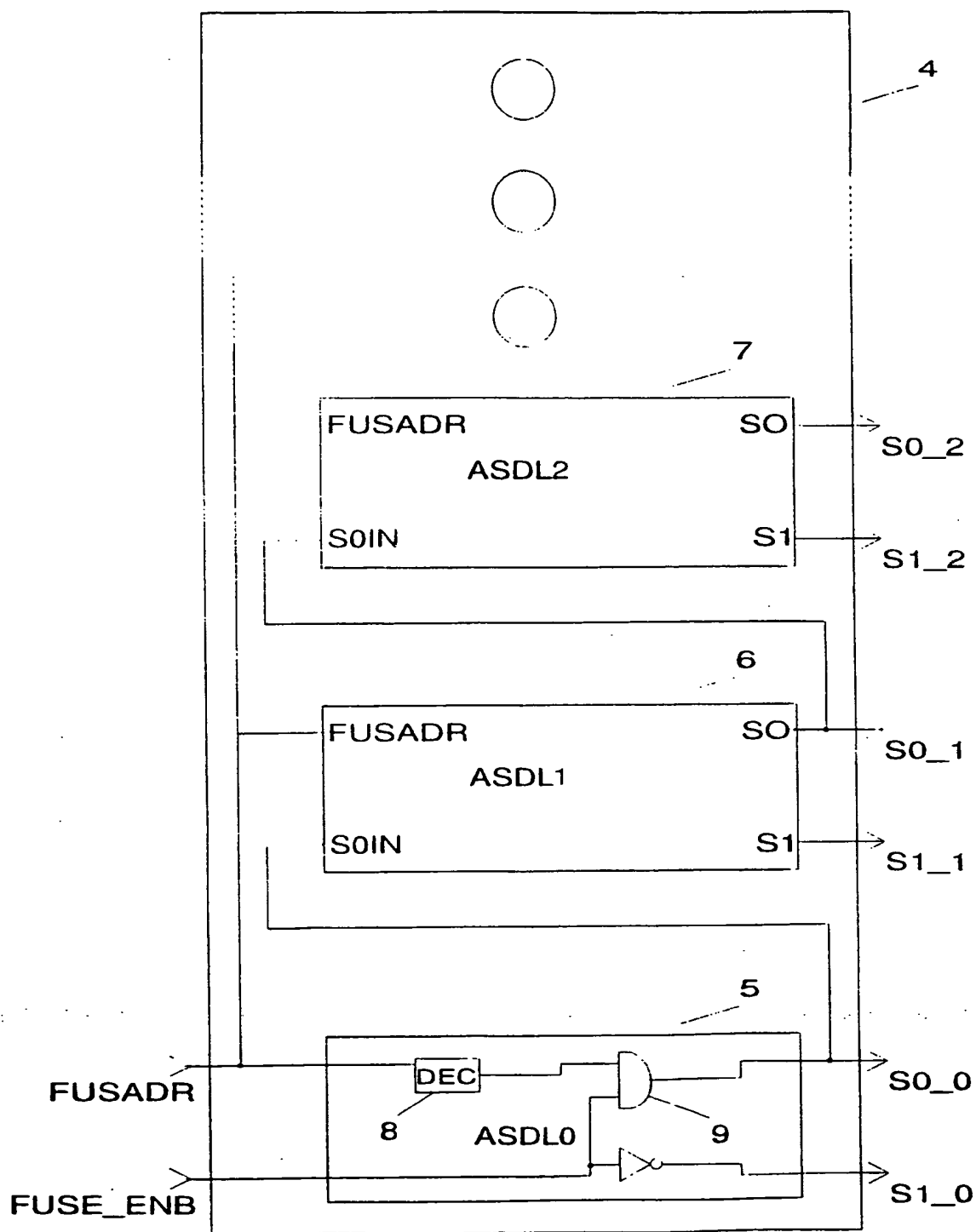


FIG. 2

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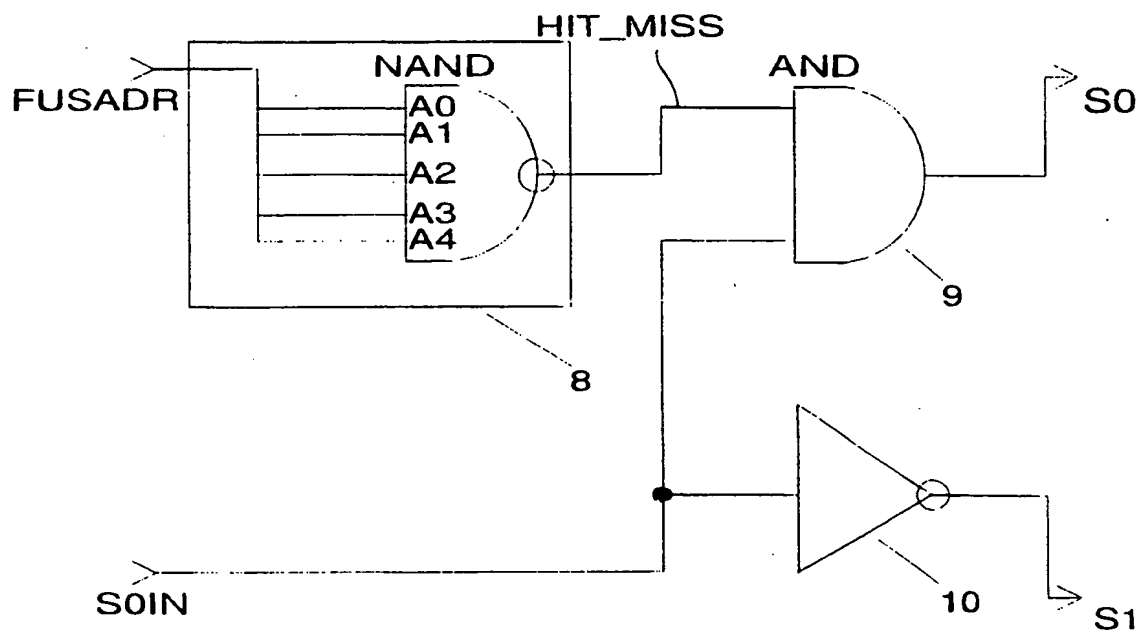
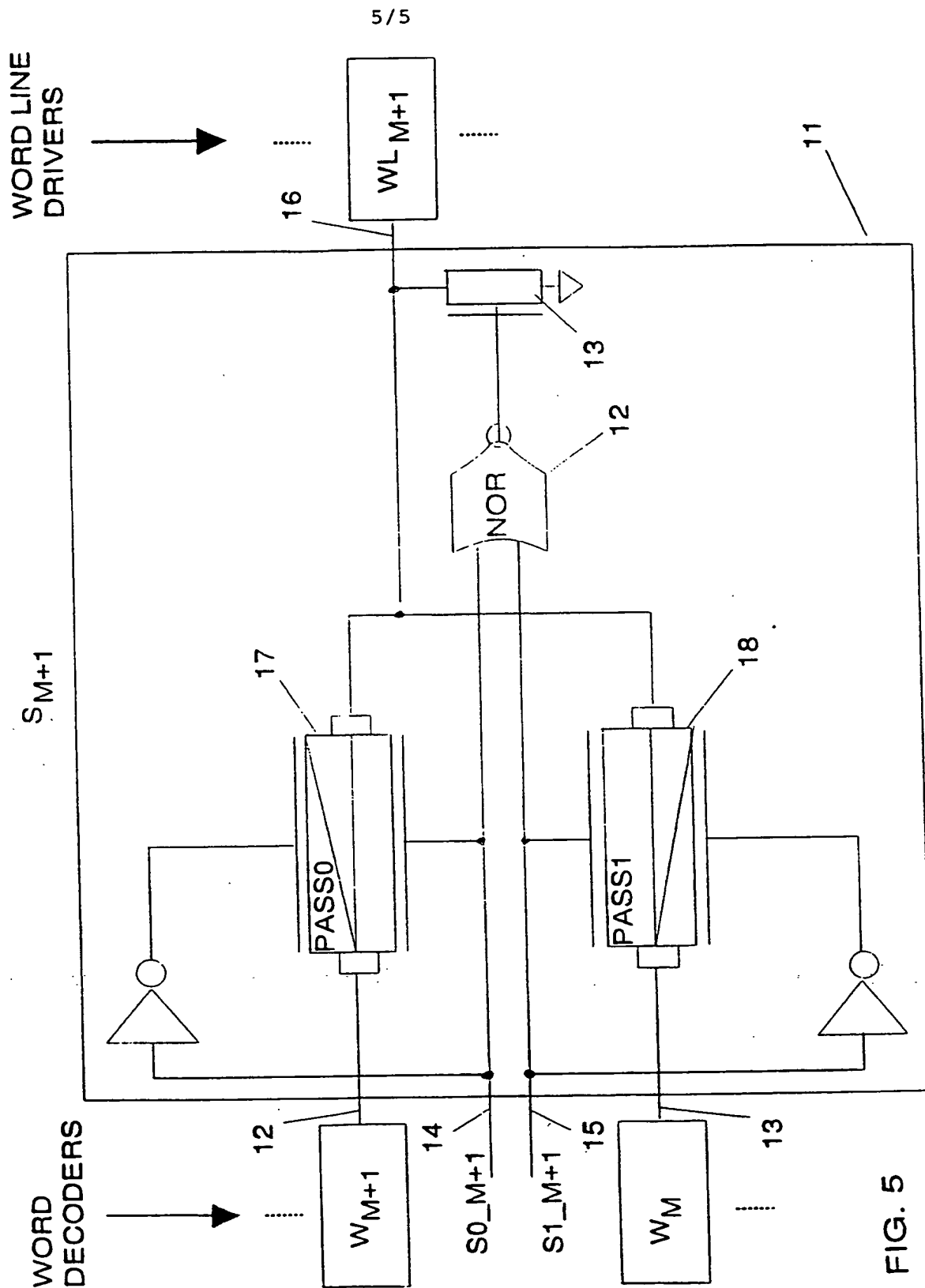


FIG. 3

FUSADR	ASDL0	ASDL1	ASDL2	ASDL3	...
B0		A0		A0	
B1			A1	A1	
B2					
B3					
B4					
$\overline{B0}$	A0		A0		...
$\overline{B1}$	A1	A1			
$\overline{B2}$	A2	A2	A2	A2	
$\overline{B3}$	A3	A3	A3	A3	
$\overline{B4}$	A4	A4	A4	A4	

FIG. 4





# INTERNATIONAL SEARCH REPORT

International Application No  
PC/EP 95/02183

**A. CLASSIFICATION OF SUBJECT MATTER**  
IPC 6 G06F11/20

According to International Patent Classification (IPC) or to both national classification and IPC

**B. FIELDS SEARCHED**

Minimum documentation searched (classification system followed by classification symbols)  
IPC 6 G06F

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

**C. DOCUMENTS CONSIDERED TO BE RELEVANT**

Category *	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
X	EP,A,0 401 957 (MITSUBISHI DENKI KABUSHIKI) 12 December 1990 see column 9, line 52 - column 12, line 38; figures 3,4,6,10,15	9
Y	see column 16, line 49 - column 17, line 18 see column 27, line 51 - column 28, line 27	1
Y	--- EP,A,0 361 404 (NEC CORPORATION) 4 April 1990 see abstract; figures 2-5	1
A	--- US,A,5 146 429 (KAWAI ET AL.) 8 September 1992 see abstract; figure 5 -----	1

☐

Further documents are listed in the continuation of box C.

☒

Patent family members are listed in annex.

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Date of the actual completion of the international search

5 March 1996

Date of mailing of the international search report

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# INTERNATIONAL SEARCH REPORT

Information on patent family members

International Application No

PCT/EP 95/02183

Patent document cited in search report	Publication date	Patent family member(s)	Publication date
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		US-A- 5471427	28-11-95
		US-A- 5134585	28-07-92
		US-A- 5379258	03-01-95
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		DE-D- 68925090	25-01-96
US-A-5146429	08-09-92	JP-A- 4090193	24-03-92

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